

PULSE WIDTH MODULATOR USING OTRA BASED TIMER CIRCUIT

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Abstract

This paper proposes a modified PWM circuit using OTRA based analog timer circuit. Designing a pulse width modulator circuit by employing the proposed analog timer circuit results in a waveform having variable duty cycle and time period. The timer circuit consists of comparators designed using OTRA and flip flops. The fabrication of timer circuit on a monolithic integrated circuit is easier as the carrier signal type is exponential and no additional circuitry is needed unlike for triangular and sawtooth pulses. The process parameters of 0.5 μm CMOS have been used for performing an intensive simulation of the proposed PWM circuit.

Keywords:

Timer Circuit, Pulse Width Modulator (PWM), OTRA, Flip-Flop

1. INTRODUCTION

Analog Signal processing is irreplaceable in today's world as all real signals are analog in nature. In fact, for digital processing also, we need analog blocks to convert analog signal to digital signals like data convertors, voltage oscillators etc. Recently current mode analog signal processing has captured attention of scientific community due to its ability to curtail voltage swings which results in an improved capability to handle signals particularly at low power supply. This makes it a viable option to employ in mixed signal design circuits functioning at 3.0 Volts supply voltage, which in near future, is expected to be a standard value for the CMOS industry.

As technology length shrinks from micron to submicron region, the supply voltages requirement has also been lowered to 1.5 Volts. At this reduced supply voltage, it is nearly impossible for a voltage mode transistor circuit to be designed which possesses a wide dynamic range and high linearity. In recent times, current mode circuits have substituted voltage mode circuits in most of the future applications due to their inherent advantages over the latter [2].

The timer circuits are being widely employed in the analog industry. Conventionally, op-amp, a voltage mode circuit is being used to design classical analog timers [3]. It has its own limitations as compared to current mode circuits. Voltage mode circuit generally has low slew rate and lower bandwidth as compared to current mode circuits. The constituents of current mode circuits like Current Conveyor, Operational Trans-conductance Amplifier, Current Differential Trans-Conductance Amplifier, Operational Trans-resistance Amplifier etc. have a wider dynamic range and bandwidth. They also have larger linearity than their voltage mode counter parts like op-amps [4].

Therefore, such functional blocks are a better alternative than op-amp for designing of timer circuits [5] [6]. The PWM circuit is extensively used to govern the speed of DC motors [11] [12], manage the power supplied to electrical devices [7]-[10], communication, RADAR detection and so on. In this proposed

paper, an OTRA based timer circuit which operates at low voltages has been employed to design pulse width modulator (PWM). Realization of active blocks is extensively discussed in Section 2 and Section 3. The methodology of proposed Pulse Width Modulator is discussed in Section 3. In section 4, we analyze and compare the features and parameters of proposed PWM with other PWMs.

1.1 ACTIVE BLOCK - OTRA

OTRA is a current controlled voltage source. It has a large bandwidth, high gain and a high slew rate which is independent of gain. It has no parasitic effects due to its virtually grounded input terminals [15]. Different CMOS structures of OTRA have been surveyed in literature [17,18]. The Fig.1 shows the basic block diagram of OTRA and its output V_o is computed using its port characteristics given in Eq.(1), where R_m is the transresistance gain of OTRA [19].

$$\begin{bmatrix} v_p \\ v_n \\ v_z \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ R_m & -R_m & 0 \end{bmatrix} \begin{bmatrix} i_p \\ i_n \\ i_z \end{bmatrix} \quad (1)$$

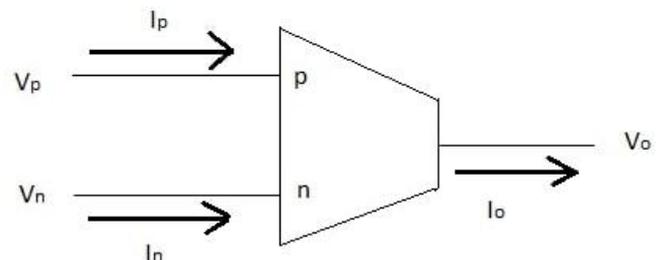


Fig.1. Block diagram of OTRA

The Fig.2 depicts the CMOS structure of OTRA [20]. The cascaded inverter [M15-M18] takes the output of this circuit which provides voltages to V_{CC} and ground respectively.

1.2 OTRA BASED TIMER CIRCUIT (CIRCUIT DESCRIPTION)

The OTRA based timer circuit is shown in Fig.3 [27]. It has 2 comparators, namely comp_1 and comp_2, S-R flip-flop, NMOS M1 and multiple resistors [21]. It has 3 input pins, namely, threshold pin, trigger pin, discharge pin, output pin (V_{out}) and ground pin. The ground pin is connected to common ground. The function of comparator is to compare the input voltage/ current with the reference voltage/current. In the current work, OTRA is used to design a comparator which compares the current input to the two terminals resulting in an output voltage which may be positive or negative. The inverter takes the output of this circuit, thereby providing voltages to V_{CC} and ground respectively.

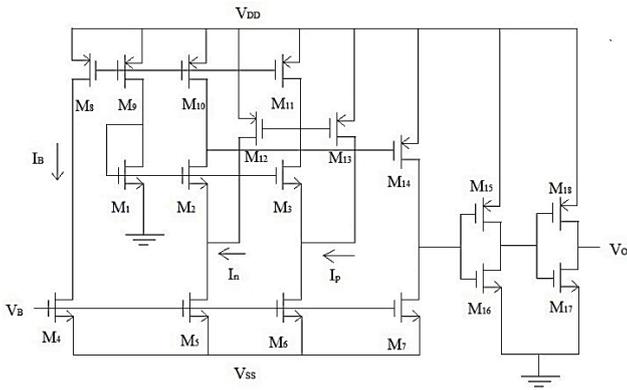


Fig.2. CMOS Realization of OTRA

2. PROPOSED PWM CIRCUIT

The Fig.4 shows the proposed PWM circuit. A modulating current signal is applied at input (I_n) of comparator₁ (comp₁). The amplitude of modulating signal is less than total current I_n . I_1 and I_2 are defined such that I_1 is the current when the amplitude of modulating signal is higher and I_2 is current when the amplitude of modulating signal is lower. I_{total} is the total current provided by the power supply.

2.1 OPERATION

Initially, the capacitor is in discharge condition, so voltage across the capacitor C_1 is zero. The current at p -terminal of comparator₁ is zero and at n -terminal of comparator₁ is the addition of $2/3 * I_{total}$ (due to current divider circuit) and modulating current signal, hence, the output of comparator₁ is zero. Also, the current at n -terminal of comparator₂ is zero (since the voltage across the capacitor C_1 is zero) and current at p -terminal of comparator₂ is $1/3 * I_{total}$ due to which output voltage of comparator₂ is V_{cc} . Output of comparators in circuit is connected to the S-R flip-flop as shown in Fig.2 (S is connected to output of Comparator₁ and R is connected to output of Comparator₂), hence, the inputs of S-R flip flop are $S=0, R=1$. The output of S-R flip-flop for $S=0, R=1$ will be $Q=0$. This output of S-R flip-flop is connected to the gate of NMOS M_1 . Due to zero gate voltage, NMOS M_1 will be in OFF condition. Capacitor C_1 starts charging towards supply voltage V_{cc} through resistors R_3 and R_5 . When capacitor C_1 gets charged to such a value that the current through n -terminal of comparator₂ becomes greater than or equal to the current through p -terminal, i.e., $1/3 * I_{total}$, output of comparator₂ becomes 0 Volt. However, the current through p -terminal of comparator₁ is still less than the addition of $2/3 * I_{total}$ and modulating signal and therefore output of comparator₁ is at 0V. Now, the inputs of S-R flip-flop will be $S=0$ and $R=0$. So, the output will remain in its previous state which means $Q=0$. The NMOS M_1 is in OFF condition due to which capacitor C_1 continues to charge towards a voltage so that the current through p -terminal of comparator₁ becomes more than addition of $2/3 * I_{total}$ and modulating signal resulting in its output to become 1. Although, the output of Comparator₂ remains 0V resulting in the S-R flip flop's inputs being $S=1, R=0$ and its output will be $Q=1$. NMOS M_1 will now be in ON condition and hence capacitor C_1 begins discharging to such a voltage level that the current through the p -terminal of comparator₁ becomes less than the

current through n -terminal (i.e., $2/3 * I_{total}$). The output of comparator₂ stays unchanged at 0 Volt. So, the flip-flop inputs are now $S=0, R=0$ resulting in an unchanged output, i.e., $Q=1$. NMOS M_1 will now be in ON condition and therefore capacitor C_1 keeps discharging to 0V through NMOS M_1 . As capacitor C_1 continues to discharge, at some point of time the current through n -terminal of comparator₂ becomes less than the p -terminal current (i.e., $1/3 * I_{total}$) due to which the output of comparator₁ stays unchanged at 0V. Now, the flip flop inputs are $S=0, R=1$, so its output is in $Q=0$ state. Therefore, NMOS M_1 turns OFF and capacitor C_1 stops discharging through the NMOS M_1 path. Subsequently, the current level at n -terminal of comparator₁ becomes equal to the sum of $2/3 * I_{total}$ and modulating current signal resulting in repetition of the same cycle. But the time required for charging and rise in voltage level with charge in capacitor is different which is lesser compared to the values when sinusoidal current was I_1 . The aforementioned process is repeated continuously with the application of a periodic pulse used for triggering. The waveform obtained at the output of proposed PWM modulator is shown in Fig.10.

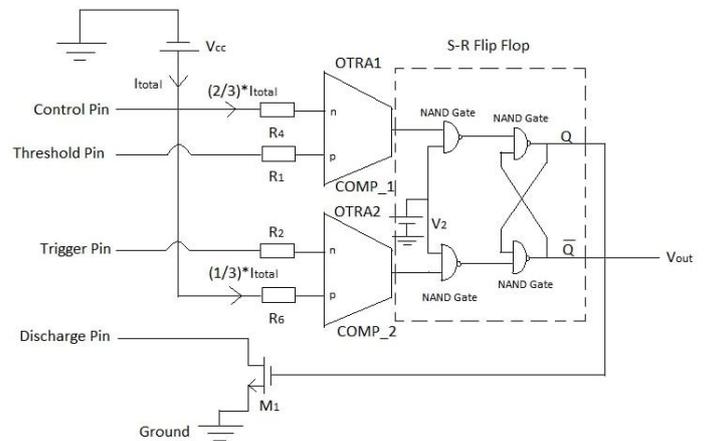


Fig.3. OTRA based timer circuit

3. SIMULATION AND EXPERIMENTAL RESULTS

This operation of the PWM modulator being proposed is established by an intensive simulation in SPICE using $.5\mu\text{m}$ CMOS technology. The timer circuit is constructed using the CMOS structure of OTRA shown in Fig.2. The parameters being used are:

- Supply voltages: $\pm 1.5\text{V}$
- W/L ratio of NMOS transistor M_1 used in timer circuit: $2\mu\text{m} / 100\mu\text{m}$
- W/L ratios of transistors used in the OTRA are $M_{1,2,3}$: $100/2.5$, $M_{4,7,12,13}$: $10/2.5$, $M_{5,6}$: $30/2.5$, $M_{8,11,14}$: $50/0.5$

The Table.1 shows the component values being used in the PWM circuit. The duty cycle is calculated using the formula given below:

$$\text{Duty Cycle} = T_{on} / (T_{on} + T_{off}) \quad (1)$$

The Fig.7 shows the voltage across capacitor C_1 and Fig.8 shows the output voltage without modulating current signal. In the current PWM circuit, a modulating signal is applied at n -terminal of comparator₁ as shown in Fig.4. Applying a modulating signal

with 10KHz frequency results in an output which is shown in Fig.9. Similarly, applying a modulating signal with 20KHz frequency results in an output which is shown in Fig.10. The Fig.11 and Fig.12 show the frequency domain representations for the modulating signal with a frequency of 5 KHz and the output signal respectively. Since the exponential voltage waveform across the capacitor is being used as carrier wave, therefore, no additional circuitry is needed for the triangular/sawtooth waves.

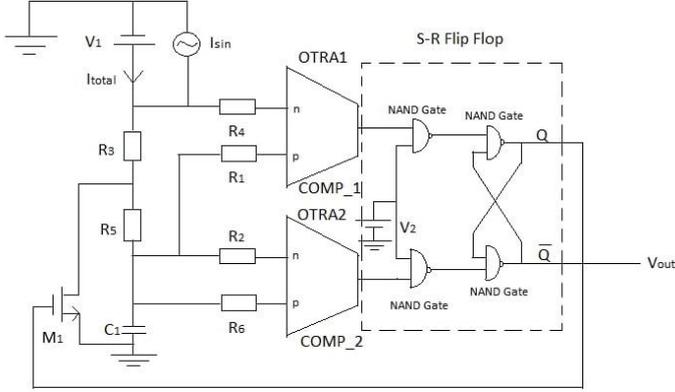


Fig.4. Proposed PWM using OTRA based timer circuit

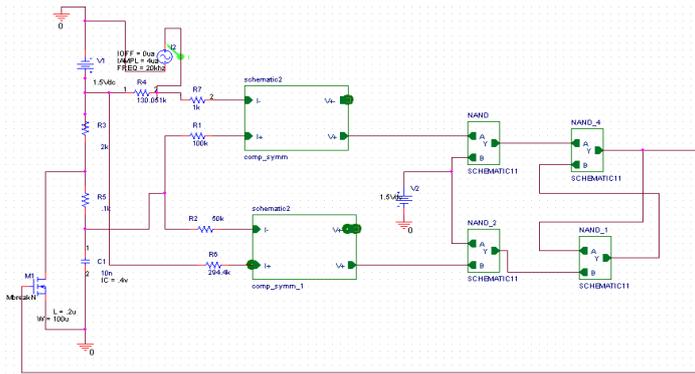


Fig.5. Spice schematic of proposed circuit

Table.1. Component values used in proposed circuit

Component	Values
R_1	100 k Ω
R_2	50 k Ω
R_3	2 k Ω
R_4	129.68 k Ω
R_5	0.1 k Ω
R_6	294.4 k Ω
R_7	1 k Ω
C_1	100 pF

Table.2. Comparison of proposed PWM circuit with existing PWM circuits

Active Components	Passive Components	Carrier Signal	Electronic Tunability
1 CC-II, 2 op-amps [22]	1 C, 3 R	Triangular	No
1 op-amp [23]	1 C, 3 R	Exponential	No
3 Operational Trans-conductance Amplifier [25]	1 C, 2 R	Triangular	Yes
4 Operational Trans-conductance Amplifier [26]	1 C	Triangular	Yes
Proposed 2 OTRAs, NAND gate-based Flip Flop	1 C, 6 R	Exponential	Yes

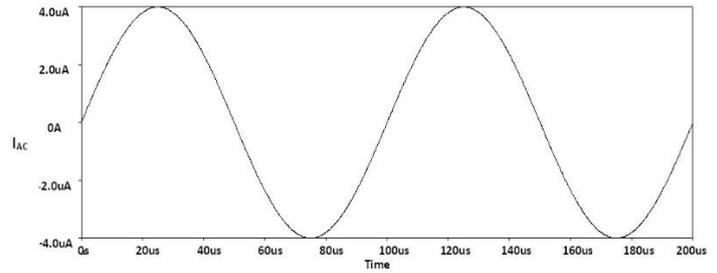


Fig.6. Modulating current signal

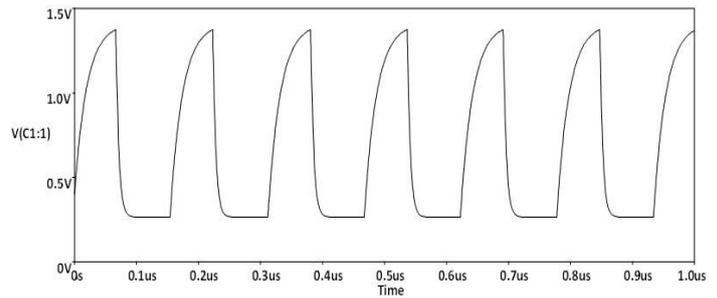


Fig.7. Voltage across capacitor without modulating signal

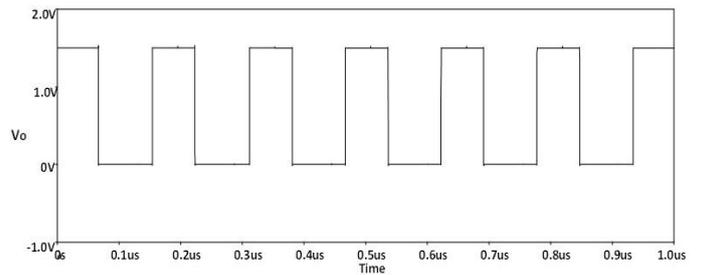


Fig.8. PWM output without modulating signal

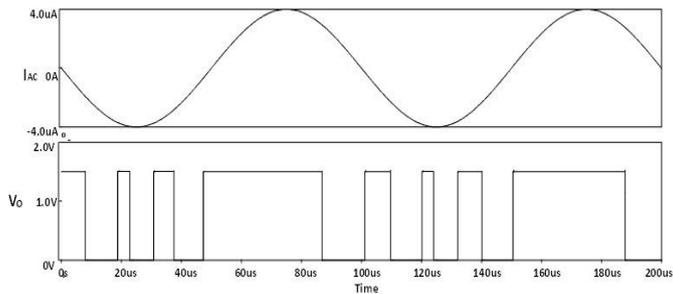


Fig.9. Output of PWM when modulating signal with 10KHz frequency is applied

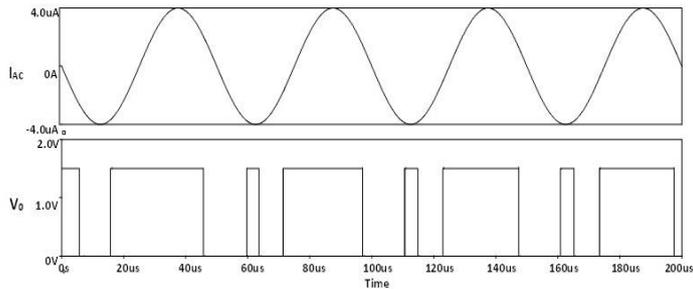


Fig.10. Output of PWM when modulating signal with 20KHz frequency is applied

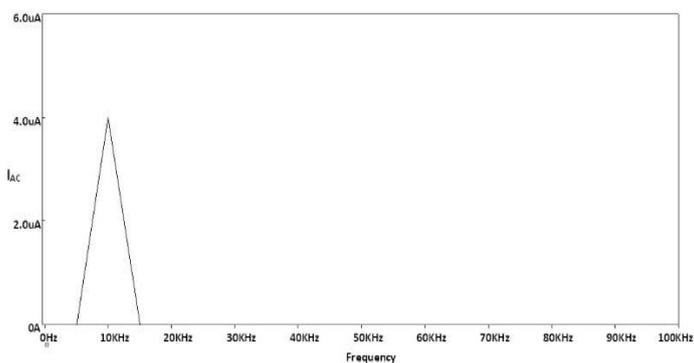


Fig.11. Frequency domain representation of modulating signal

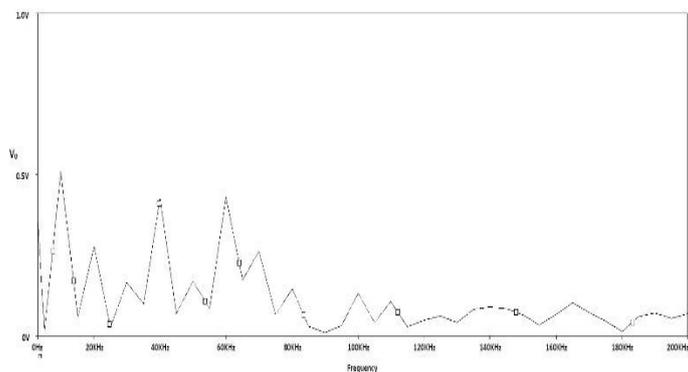


Fig.12. Frequency domain representation of output signal

4. COMPARISON

When the proposed work is compared with existing analog pulse width modulator circuits [22]-[26], it is observed that in existing circuits the PWM output is being generated by making a

comparison between the reference sawtooth or triangular waveform and the modulating signal. It is also observed in [23] that PWM circuit is designed using op-amp, which is a voltage mode circuit. Voltage mode circuits have their own limitations as compared to current mode circuits. Voltage mode circuit generally has low slew rate and lower bandwidth as compared to current mode circuits.

The proposed circuit is based on current mode circuit, so that the performance does not get degraded for high frequency applications due to limited slew rate and gain-bandwidth product. A detailed comparison of the above fact is shown in Table 2. In addition to this OTRA input terminals are also virtually grounded due to which there are no parasitic effects on the proposed circuit.

5. CONCLUSION

This work presents a PWM generator using OTRA based timer circuit by utilizing the non-linear properties of OTRA. It is observed that the proposed circuit performs better as compared to the existing circuits for high frequency switching applications.

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