DESIGN AND ANALYSIS OF COMPACT PRINTED TRIPLE BAND ANTENNA FOR WI-MAX, WLAN AND WIRELESS SYSTEMS

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Abstract

In this paper, we design a compact single antenna which is used for different bands. The term compact here means that the size of the antenna is very small to the tune of just $24 \times 30 \text{mm}^2$. This antenna is fabricated on roger RTduroid 5880 with thickness 0.79 mm, and is designed using partial ground plane. The antenna is equipped with two inverted L shaped stub and a stub of combination of inverted L shape and T shape; and it resonates at three frequencies, 2.57GHz, 3.52GHz, and 5.51GHz. As the operating range of antenna covers these frequencies, its main use is in Wi-MAX (wide interoperability for microwave access) and WLAN (wireless local area network). All the parameters and details are elaborated by the ensuing parts of this paper.

Keywords:

Compact Size Antenna, Single Antenna, RTduroid, Inverted L and T Shape

1. INTRODUCTION

There is high demand for small-sized equipment in the electronics and wireless equipment market, because the small-sized components occupy less space giving the scope for the inclusion many latest features, and for effective heat sinking. Reducing the size of the antenna used in a wireless system is of key importance from the perspective reducing the size of the entire system itself. The antenna discussed in this study is mainly designed for use in Wi-MAX and WLAN systems.

For providing a perfect antenna for this requirement we studied numerous published papers to get the knowledge of design and measurements of antenna. A number of multiband antennas have been reported by researchers having good efficiency, CPW-fed stub loaded slot antenna is design in reference 1 with dimension of $0.245\lambda_o \times 0.53\lambda_o \times 0.013\lambda_o$; a symmetrical U shaped antenna is designed in reference 2 with dimension of $0.166\lambda_o \times 0.32\lambda_o \times 0.006\lambda_o$, reference 3 represents the F shape stub antenna with dimension of $0.187\lambda_o \times 0.294\lambda_o \times 0.008\lambda_o$, so many more references are given from which we are taking help to design our proposed antenna.

In our work we are only mentioning the main references [1]-[12]. All the researchers of these papers have already created multiband antenna using different dimensions and material, so we are taking the best features from their antenna to create an optimized proposed antenna. In our article, we design a compact printed antenna. It consists three inverted L stub with different height (vertical) but with same thickness on Roger RTduroid 5880 dielectric substrate and the middle inverted L stub connected with a T stub (90 degree rotated T stub). Due to the varied heights of the inverted L stub (and T stub) they resonate at different frequencies. This feature is very useful for us to use a single antenna for different frequency bands.

| Table.1. Troposed work with reference and mas | | | | | | | | |
|---|--|---|-------------------------------------|--|--|--|--|--|
| No. of bands in antenna | Dimensions of antenna | Bandwidth (GHz) S ₁₁ ≤10dB | Realized gain | | | | | |
| Triple band [1] | $\frac{30 \times 65 \times 1.6}{(0.245 \lambda_o \times 0.53 \lambda_o \times 0.013 \lambda_o)}$ | 2.375-2.525 3.075-3.8 5.0-6.9 | 1.1-1.5 4.6-5.6 2.0-3.6 | | | | | |
| Triple band [2] | $\begin{array}{c} 20 \times 38.5 \times 0.8 \\ (0.166 \lambda_o \times 0.32 \lambda_o \times 0.006 \ \lambda_o) \end{array}$ | 2.10-2.49 3.22-4.30 4.89-6.12 | 2.7-3.2 3.1-3.5 2.8-3.3 | | | | | |
| Triple band [3] | $23 \times 36 \times 1 \ (0.187 \lambda_o \times 0.294 \lambda_o \times 0.008 \ \lambda_o)$ | 2.35-2.59 3.31-3.93 5.07-6.35 | 2.67-3.19 2.05-2.24 2.66-3.48 | | | | | |
| Triple band [4] | $\begin{array}{c} 28 \times 32 \times 1 \\ (0.23 \lambda_o \times 0.266 \lambda_o \times 0.008 \lambda_o) \end{array}$ | 2.29-2.88 3.26-3.88 4.17-6.07 | 3.8-4.4 4-4.65 1.9-3.5 | | | | | |
| Triple band [5] | $33 \times 50.9 \times 0.8 \\ (0.319 \lambda_o \times 0.492 \lambda_o \times 0.007 \lambda_o)$ | 2.8-3.0 3.3-3.5 3.8-8 | 2.32 1.21 -6 | | | | | |
| Triple band [6] | $\begin{array}{c} 26 \times 30 \times 0.8 \\ (0.212 \lambda_o \times 0.25 \lambda_o \times 0.006 \lambda_o) \end{array}$ | 2.33-2.55 3.0-3.88 5.15-5.9 | 1.08-1.39 2.35-3.49 2.46-3.28 | | | | | |
| Triple band [7] | $\begin{array}{c} 23 \times 38 \times 1.6 \\ (0.184 \lambda_o \times 0.304 \lambda_o \times 0.013 \lambda_o) \end{array}$ | 2.28-2.56 3.29-4.21 5.05-5.91 | 1.48-1.96 2.1-3.22 2.63-3.56 | | | | | |
| Triple band [8] | $\begin{array}{c} 40 \times 40 \times 0.8 \\ (0.325 \lambda_o \times 0.325 \lambda_o \times 0.007 \lambda_o) \end{array}$ | 2.35-2.58 3.25-4 4.95-5.9 | -0.3 0.9 3.8 | | | | | |
| Triple band [9] | $40 \times 45 \times 1.6$ (0.28 $\lambda_o \times 0.31 \lambda_o \times 0.011 \lambda_o$) | 2.02-2.14 4.26-4.28 5.45-5.56 | 1.87 2.90 4.13 | | | | | |
| Triple band [10] | $37 \times 40 \times 1.58$ (0.302 $\lambda_o \times 0.327\lambda_o \times 0.013\lambda_o$) | 2.30-2.75 3.19-3.18 5.06-6.15 | 2.5 0.824 2.02 | | | | | |
| Triple band [11] - target antenna | $24 \times 30 \times 0.79$ $(0.2\lambda_o \times 0.25\lambda_o \times 0.006\lambda_o)$ | 2.43-2.67 3.37-3.58 5.1-5.65 | 0.5-1.75 1.3-1.9 1.7-1.9 | | | | | |

Table.1. Proposed work with reference antennas

It is a triple layer antenna with partial ground plane which means the bottom layer does not cover the whole dielectric substrate. In this antenna design we are capable of changing the resonating frequency by changing the thickness (horizontal) of stub. We fabricate and design the antenna for the purpose of validation. This compact printed multiband antenna works for Wi-MAX at 2.5GHz, 3.5GHz, 5.5GHz and also 5.2GHz and 5.5GHz for WLAN system. This compact printed multiband antenna works for Wi-MAX at 2.5GHz, 3.5GHz, 5.5GHz and also 5.2GHz and 5.5GHz for WLAN system.

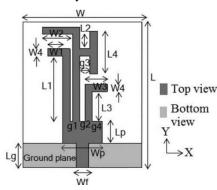


Fig.1 Geometry of proposed compact printed multiband antenna

2. CONFIGURATION AND ANALYSIS OF ANTENNA

Above Fig.1 represents the geometric view of the proposed antenna in *xy* plane. First we have to design the software version of the antenna in CST. All the dimension of the antenna is provided in the Table.2.

Table.2. Parameter of proposed antenna

| Parameters | L | W | L_p | L_g | W_{f} | W_p | L_1 | L_2 | L ₃ |
|-------------|-------|-------|-------|-------|---------|-------|-------|------------|----------------|
| Values (mm) | 30 | 24 | 4.5 | 5 | 2.44 | 8 | 13.5 | 3 | 6.15 |
| Parameters | L_4 | W_1 | W_2 | W_3 | W_4 | g_1 | g_2 | g 3 | g 4 |
| Values (mm) | 9.1 | 3 | 6 | 4.5 | 1.5 | 0.5 | 1 | 2 | 2 |

The antenna is designed on low loss dielectric Roger RTduroid 5880 substrate with partial ground plane. Thickness of dielectric substrate is 0.79mm and value of dielectric constant is 2.2 and loss tangent is 0.0009.

The antenna is designed using a 50 Ω micro strip feed line with partial ground plane and two inverted L shaped stub and a mix inverted L and T stub. The overall size of the antenna is 24×30mm². The optimized antenna's dimensions are given in Table.2. In the designing process of this antenna, we are using four prototypes antenna as step by step design. First, we have to design a rectangular patch dimension $L_p \times W_p$ and excited through feed line (50 Ω) of width W_f . This rectangular patch resonates at higher frequencies. For reducing resonance frequency, we added different stubs in it. An inverted L-stub is added to the left upper side of the rectangular patch and we call it Resonator 1. Now, it provides resonance at 3.56GHz, whereas our purpose is to design the antenna to be used at 2.5/3.5/5.5GHz for Wi-MAX and 5.2/5.5GHz for WLAN.

Again we are added a bigger inverted L-stub to the rectangular patch at a gap of g_1 from previous stub. It is called Resonator 2. Now the antenna is found to resonate at 2.57GHz and 3.5GHz. We, then, added a horizontal T-stub to the largest inverted L-stub. It is called Resonator 3. Due to T-stub, the lower frequency reduces and second resonance remains same. Both Resonator 2 and Resonator 3 are called dual band antenna because they are resonating at two different frequencies.

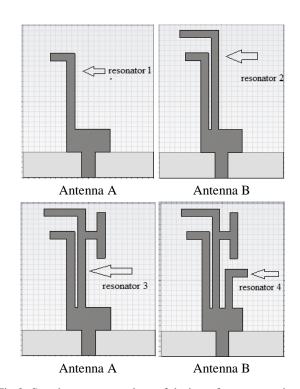


Fig.2. Step by step procedure of design of compact printed multiband antenna

Now we add the smallest inverted L-stub to the rectangular patch at a gap of g_2 from the largest inverted L-stub. It is called Resonator 4 and this whole setup is our final antenna. For getting the further result, we have to simulate the antenna. After simulation we plot a graph comparison between real impedance and imaginary impedance as seen in Fig.2. As displayed in Fig.4, we conclude that the three encircled regions define the three resonance mode in the proposed antenna, namely 2.57, 3.52 and 5.51GHz.

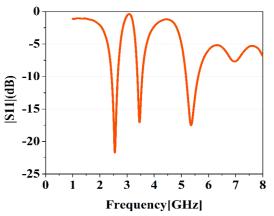


Fig.3. Reflection coefficient ($|S_{11}| dB$) against frequency

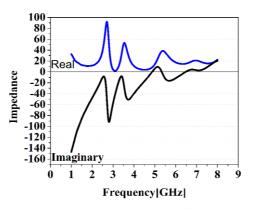


Fig.4. Input impedance against frequency

To understand the antenna's working and resonating phenomenon, we observed the current distribution of the antenna (given in Fig.7). We resonated the antenna at three frequencies which is 2.57, 3.52 and 5.51GHz and defined the role of each stub for their frequency resonance.

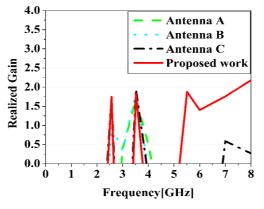


Fig.5. Realized gain against antenna configuration

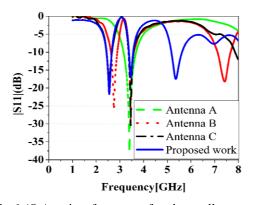


Fig.6. $|S_{11}|$ against frequency for above all antenna

First we resonated the proposed antenna at 2.57GHz and observed the surface current distribution is mainly in the Resonator-3, which is the combination of inverted L-stub and horizontal T-stub. So the resonance frequency 2.57GHz occurred due to the Resonator-3. It is shown in Fig.7(a).

Again we observed the surface current distribution in proposed antenna at second resonance frequency i.e. 3.52GHz. At this resonance frequency, most of surface current flowed in the left most inverted L-stub in the antenna. So, we conclude that the second resonating frequency (3.52GHz) occurred due to left most

inverted L-stub as shown in Fig.7(b). Again we observed the surface current distribution in the proposed antenna at third resonance frequency i.e. 5.51GHz. At this frequency most of surface current flowed in the right most inverted L-stub. So, we concluded that the third resonating frequency occurred due to the right most inverted L-stub of the proposed antenna.

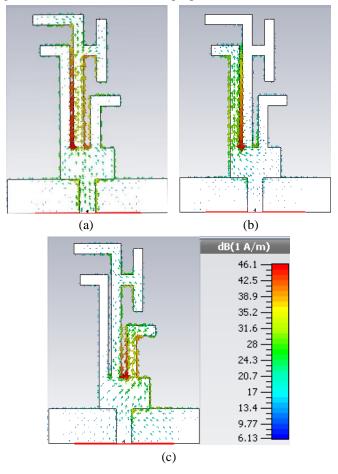


Fig.7. Surface current distributions at frequency (a) 2.57GHz (b) 3.52GHz (c) 5.51GHz

3. DESIGN PROCEDURE FOR PROPOSED ANTENNA

From the surface current distribution phenomenon discussed above, we understand that different resonance frequencies occurred due to different resonators present in the proposed antenna. For the design we need to know the f_{rn} (n^{th} resonance frequency) and ε_{eff} (effective dielectric constant), given as below:

$$f_{rn} = \frac{c}{4L_{sn}\sqrt{\varepsilon_{eff}}}; n = 1, 2, 3$$
(1)

$$\varepsilon_{eff} \approx \frac{\varepsilon_r + 1}{2}$$
 (2)

where ε_r = dielectric constant of substrate, ε_r = 2.2 (for Roger RT/duroid dielectric), L_{sn} = complete length of resonator which relate to f_{rn} , λ_g = guided wavelength.

As per the aforementioned discussion, we come to know that the first resonance frequency occurred due to resonator-3 which included inverted L-stub and horizontal T-stub. So, the complete length of resonator-3 is approximated as $W_2+g_3+L_1+L_2$. Length is always equal to one fourth of guided wavelength in medium given as,

$$L_{sn} = L_2 + W_2 + g_3 + L_1 \approx \lambda_g / 4$$
 (3)

All the values are already optimized and calculated in Table.1.

$$L_{sn} = 24.5 \text{mm}$$
 $\varepsilon_{eff} = 1.6$

From these values we calculate the first resonance with the help of equation 1 is $f_{r1} = 2.42$ GHz. This is very close to the simulated resonance frequency 2.57GHz. Using the same procedure mentioned above, we calculated the second and third resonating frequencies. So, as the resonator 1 provided the second resonance frequency, the complete length of resonator 1 is given as $L_1 + W_1$.

$$Ls_2 = W_1 + L_1 \approx \lambda_g / 4 \tag{4}$$

Applying these values from Table.1 into Eq.(1), we find that $L_{s2} = 16.5$ mm and $f_{r2} = 3.59$ GHz, which is very close to the simulated resonance frequency 3.52GHz. Now calculating these values or the third resonance frequency which occurred due to resonator-4, the complete length of the resonator-4 is calculated as L_3+W_3 , given as below,

$$Ls_3 = W_3 + L_3 \approx \lambda_g/4 \tag{5}$$

All the values collected from Table.1 are applied to the aforementioned formulas to get, $L_{s3} = 10.65$ mm and $f_{r3} = 5.57$ GHz, which is closer to the simulated resonance frequency 5.51GHz. So, the antenna design justifies both the practical and theoretical approach.

| Name of antenna | Pattern of antenna | Bandwidt h S11 ≤10 GHz | Realized gain S11 ≤10 | No. of bands |
|--------------------|--|------------------------------------|--------------------------------|-----------------|
| Antenna 1 | Left side inverted L- stub | 3.1-3.72 | 0.55-1.67 | Single |
| Antenna 2 | Left and middle inverted L-stub | 2.62-2.92 3.37-3.6 | 0.5-1 1.35-1.83 | Double |
| Antenna 3 | Antenna 2 with horizontal L-stub | 2.42-2.68 3.35-3.6 | 0.2-1.7 0.1-2 | Double |
| Desired antenna | Antenna 3 with right most Inverted L- stub | 2.43-2.67 3.37-3.58 5.1-5.65 | 0.5-1.75 1.3-1.9 1.7-1.9 | Triple |

Table.3. Comparison of antenna factors

4. PARAMETER VARIATION STUDY ON DESIGNED ANTENNA

To learn the behavior of the designed antenna we had to change some of its dimensions. We know that our antenna provides three resonance frequencies and every resonance frequency occurred due to a particular resonator present in the antenna. So, we are going to change the horizontal length of resonators, while keeping the other parameters same as before. We have seen that resonance frequency 2.57GHz is provided by resonator-3 and 3.52GHz is provided by resonator-1and 5.51GHz is provided by resonator-4. So, we are going to study the variation in W_1 of resonator-1, W_2 of resonator-3, W_3 of resonator-4. In this study of varied lengths, we increase and decrease the length by 2mm and get the result.

First we discuss the variation in W_1 . W_1 for the proposed work is 3mm and now we study the behavior of antenna at $W_1 = 1$ mm and $W_1 = 5$ mm. For $W_1 = 1$ mm and $W_1 = 5$ mm, we got the reflection coefficient graph against frequency as shown in Fig.8. It is observed from the graph that the second resonance shifted toward right for decreased W_1 and shifted toward left for increased W_1 .

Similarly, we discuss the variation in W_2 for resonator-3. W_2 for proposed work is 6mm, so we are going to study the variation at $W_2 = 4$ mm and 8mm. when the length of resonator-3 decreased to 4mm then the first resonance shifted toward right or higher frequency and for increased W_2 (8mm) second resonance shifted toward left or lower frequency side (as shown in above Fig.9).

The same procedure was repeated for variation in W_3 for resonator-4. W_3 for proposed work is 4.5mm and we are going to study the variation at $W_3 = 2.5$ mm and $W_3 = 6.5$ mm. When the length of resonator-4 decreased to 2.5mm the third resonance shifted toward right side or higher frequency side whereas for increasing in length of W_3 to 6.5mm, the third resonance shifted toward left or lower frequency side (as shown in Fig.10). We also observed that W3 is most sensitive in comparison to W_1 and W_2 .

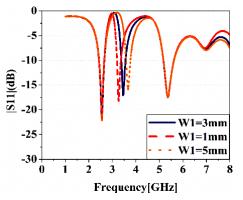


Fig.8. $|S_{11}|$ against frequency (with variation in W_1)

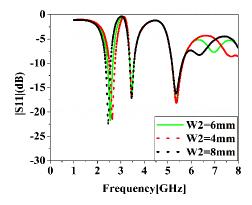


Fig.9. $|S_{11}|$ against frequency (with variation in W_2)

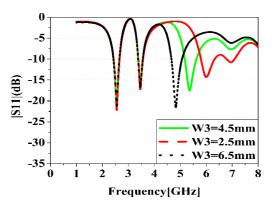


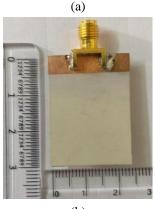
Fig.10. $|S_{11}|$ against frequency (with variation in W_3)

5. RESULTS

After designing and simulating the proposed antenna in CST microwave studio, we inferred the following results:

- The proposed antenna radiates at three frequencies (2.57GHz, 3.52GHz, and 5.5GHz). It is called as triple band antenna used for Wi-MAX and WLAN. It provides better realized gain at these resonance frequencies.
- In the simulation process of this antenna we got some important results i.e. $|S_{11}|$ against frequency, VSWR against frequency, radiation pattern against frequency, input impedance against frequency, realized gain. The antenna return loss (S_{11}) is measured on ENA series vector network analyzer, set up as shown in Fig.11(c).
- There are some variations in the measured and simulated (CST) result due to environment effects, soldering joints of SMA connector, proper material properties variation and other effects.







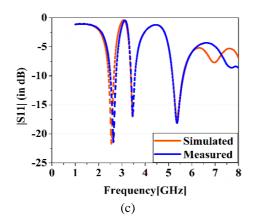


Fig.11. Proposed antenna images (a) Front view (b) Back view (c) S_{11} measured on VNA setup

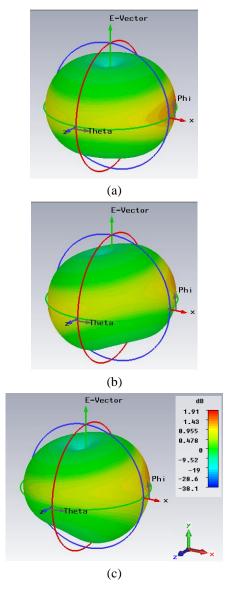


Fig.12. 3D-Radiation pattern of proposed antenna at resonance frequencies (a) 2.57GHz (b) 3.52GHz (c) 5.51GHz

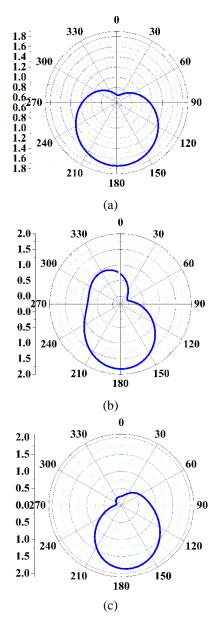


Fig.13. Radiation pattern of proposed antenna at resonance frequencies (a) 2.57GHz (b) 3.52GHz (c) 5.51GHz

6. CONCLUSION

We successfully designed the proposed multiband antenna and also simulated it on CST microwave studio and got three resonance frequencies - 2.57GHz, 3.52GHz, and 5.51GHz. So, we call it as multiband antenna. The dimension of the proposed antenna is $24\times30\times0.79$ ($0.2\lambda_o\times0.25\lambda_o\times0.006\lambda_o$) which is comparatively smaller to other antennas referenced for the purpose of this study.

Every stub in this proposed antenna is responsible for the achieved resonance frequencies. Resonator-1 is responsible for resonance frequency 3.52GHz, Resonator-3 responsible for resonance frequency 2.57GHz, and Resonator-4 is responsible for the resonance frequency 5.51GHz. This antenna has lots of

advantages like multiband, easy design, compact size, and omnidirectional pattern; and these features elevate this antenna to be a preferred component for Wi-MAX and WLAN equipment.

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