

ABSORPTIVE OPTICAL BISTABILITY IN HYBRID LASER STRUCTURE

Tan Chee Leong¹ and Pukhraj Vaya²

¹School of Photonics Science, Gwangju Institute of Science and Technology (GIST), 261 Cheomdan-gwagi-ro (Oryong-dong), Buk-gu, Gwangju, 500-712, Republic of Korea

E-mail: ccheelong@yahoo.com

²Department of Electronics and Communication, Amrita School of Engineering, Kasavanahalli, Bangalore, India

E-mail: pukhrajvaya@yahoo.com

Abstract

A hybrid bistable semiconductor laser structure consisting of an edge-emitting laser as a gain region and a vertical cavity laser as an absorber region, which translates lateral emission to vertical emission is analyzed. The device is modeled using appropriately modified rate equations and simulated using PSPICE circuit simulator. The dc sweep exhibits bistability. The hysteresis width of the bistable characteristics is found to depend on the reverse dc sweep, source resistance, and temperature. It is also noticed that the threshold currents and the hysteresis width can be controlled by the physical dimensions of gain and absorber regions and the driver circuit parameters. Turn-on delay decreases as the input current is increased.

Keywords:

Optical bistability, Edge emitting laser, Vertical cavity laser, Threshold current, Hysteresis width

1. INTRODUCTION

Bistable semiconductor laser (BSL) is one of the key devices used in optical communication, optical switching and optical memory because it provides inherent advantages such as high optical gain, low optical switching power and high switching speed. In the past one decade, many BLS switching devices such as bipolar cascade bistable, Integrated VCSEL with MESFET and VCSEL cascade-able switching devices are developed [1, 2]. These BSLs can switch either laterally or vertically. Here, we propose a hybrid bistable semiconductor laser (HBSL) structure consisting of an edge emitting laser as gain region and vertical cavity laser (VCSEL) as an absorber region which changes lateral emission to vertical emission for many possible applications. Circuit modeling technique, which can incorporate source impedance, package parasitics and device circuit interactions, is adopted to analyze the proposed device [3]. The circuit is simulated using the circuit simulator PSPICE.

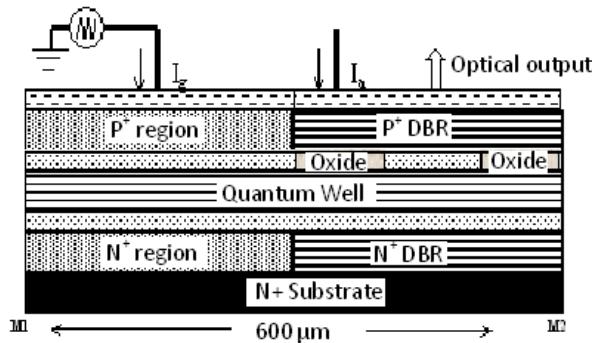


Fig.1. Proposed hybrid semiconductor laser structure for absorptive optical bi-stability

2. HYBRID BISTABLE SEMICONDUCTOR LASER STRUCTURE

The proposed HBSL structure consists of an edge emitting quantum well laser (QWL) as a gain region and a vertical cavity surface emitting laser (VCSEL) as an absorber region as shown in Fig. 1. The QWL has three InGaAs-AlGaAs strained quantum wells sandwiched between two undoped AlGaAs SCH layers. The gain region is 300 μm long while the cavity length of QWL is 600 μm, the ridge width is 2.5μm, and the width of quantum well and the barrier is 6 nm each. VCSEL and QWL share the same quantum wells. The effective aperture of VCSEL for optical output is taken as 2.5μm x 2.5μm. The active region in VCSEL is sandwiched by AlAs-GaAs DBR mirrors with reflectivity of 0.98.

3. RATE EQUATIONS

The optical fields in QWL and in VCSEL are assumed to be oscillating in TE and TM mode respectively. The effect of carrier interchange in SCH region of the QWL is incorporated in the rate equations of [4, 5]. The optical output of the HBSL is a function of the carrier concentration in absorber region, which depends on the operating conditions of both gain and absorber regions. Simple thermal equivalent circuit is added to the rate equations by making the assumption that the temperature effect in VCSEL is much larger than that in QWL. Hence the temperature effect such as leakage current in QWL region is neglected. The modified rate equations are obtained as:

$$\frac{dN_{gs}}{dt} = \frac{I_g}{qV_{gs}} - \frac{N_{gs}}{\tau_{gs}} - \frac{N_{gs}}{\tau_{gn}} \quad (1)$$

$$N_w \frac{dN_{gQ}}{dt} = \alpha_g \frac{N_{gs}}{\tau_{gn}} - N_w \frac{N_{gQ}}{\tau_{gn}} - N_w G_g (N_{gQ})(1 - \epsilon P) \quad (2)$$

$$\frac{dN_{as}}{dt} = \frac{I_a}{qV_{as}} - \frac{N_{as}}{\tau_{as}} - I_{Las} + \alpha_1 \frac{N_{aQ}}{\tau_{ate}} \quad (3)$$

$$N_w \frac{dN_{aQ}}{dt} = \alpha_a \frac{N_{as}}{\tau_{as}} - I_{LaQ} - N_w \frac{N_{aQ}}{\tau_{an}} - \alpha_1 \frac{N_{aQ}}{\tau_{ate}} - N_w G_a (N_{aQ})(1 - \epsilon P) \quad (4)$$

$$\frac{dT_a}{dt} = N_w G_g (N_{gQ})(1 - \epsilon P) + N_w G_a (N_{aQ})(1 - \epsilon P) + \Gamma \beta N_w \left(\frac{N_{gQ}}{\tau_{gn}} + \frac{N_{aQ}}{\tau_{an}} \right) - \frac{S}{\tau_p} \quad (5)$$

$$T = T_D + (IV - P_{hv})R_{th} - \tau_{th} \frac{dT}{dt} \quad (6)$$

where, I , N and P are the current applied, the carrier density and the photon density respectively. The subscripts g , a , s and Q are used for gain region, absorber region, SCH region and QW region respectively. The gain is modeled as a linear function of carrier density [3]. The photon lifetime is given by $\tau_p = (v_g(\alpha_m + \alpha_{int}))^{-1}$, where $\alpha_m (= \frac{1}{L_c} \ln \frac{1}{R})$ is the mirror loss and α_{int} in the internal loss. Transport time constant across SCH region is taken as $\frac{L_s^2}{2D_a}$. The equivalent thermal circuit is modeled based on equation (6). Leakage current, differential gain of absorber region g_a , and transparency carrier density N_{aQ} are sensitive to temperature [2, 6].

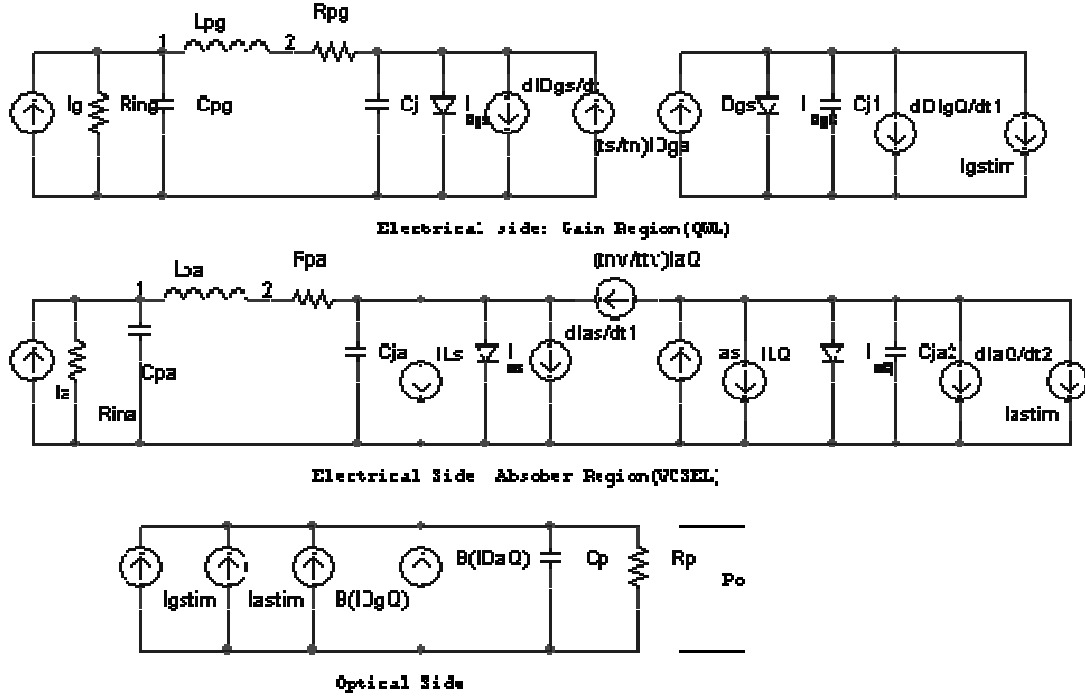


Fig.2. Equivalent electrical circuit of hybrid bistable semiconductor laser

$$I_L = I_D * C_{leak} (\exp(T/T_{ol}) - 1) \quad (7)$$

$$g_a(T) = g_a \exp[-T/T_{oa}] \quad (8)$$

$$N_{aQ}(T) = N_{aQ} \exp[T/T_{oa}] \quad (9)$$

Multiplying equations (1), (2), (3) and (4-5) by qV_{gs} , qV_{gQ} , qV_{as} and qV_{aQ} respectively, we can obtain the following expressions:

$$I_g = \tau_{gs} \frac{dI_{Dgs}}{dt} + I_{Dgs} - I_{Dgs} \frac{\tau_{gs}}{\tau_{gn}} \quad (10)$$

$$\alpha_g I_{Dgs} = \tau_{gs} \frac{dI_{DgQ}}{dt} + I_{DgQ} + I_{gstim} \quad (11)$$

$$I_a = \tau_{as} \frac{dI_{Das}}{dt} + I_{Das} + I_{Ls} + I_{DaQ} \frac{\tau_{an}}{\tau_{atte}} \quad (12)$$

$$\alpha_a I_{Das} = \tau_{as} \frac{dI_{DaQ}}{dt} + I_{DaQ} + I_{LQ} + I_{DaQ} \frac{\tau_{an}}{\tau_{atte}} + I_{astim} \quad (13)$$

$$C_p \frac{dP_o}{dt} = I_{gstim} + I_{astim} + \Gamma \beta (I_{DgQ} + I_{DaQ}) - \frac{P_o}{R_p} \quad (14)$$

where

$$I_{Dgs} = \frac{qV_{gs}N_{gs}}{\tau_{gs}}, \quad I_{Das} = \frac{qV_{as}N_{as}}{\tau_{as}},$$

$$I_{DgQ} = \frac{N_w q V_{gQ} N_{gQ}}{\tau_{gn}}, \quad I_{DaQ} = \frac{N_w q V_{aQ} N_{aQ}}{\tau_{an}}$$

$$I_{gstim} = N_w q V_{gQ} g_g (N_{gQ} - N_{og})(1 - \epsilon PS_c) PS_c$$

$$I_{astim} = N_w q V_{aQ} g_a(T) (N_{aQ} - N_{oa}(T))(1 - \epsilon PS_c) PS_c$$

The optical storage and losses are modeled as $C_p = qV_{gQ} S_c$ and

$R_p = \frac{\tau_p}{C_p}$ respectively. The large signal electrical equivalent

circuit is obtained by combining equations (10)-(14) following the scheme proposed by Ganesh et al [2] as shown in Fig 2. The source resistance and package parasitics are also included in the equivalent circuit. The optical output is assumed to be equivalent to the output voltage, P_o , across R_p . The parameter descriptions and values used in the presented study are given in Table.1.

4. SIMULATION RESULTS

The circuit model shown in Fig 2 is used to study the dc sweep and transient behavior of the HBSL. Fig 3 shows the bistable behavior of the hybrid structure. The hysteresis width of 11mA is obtained when $I_a = 10$ mA is applied to the absorber regions and I_g is varied from 0-80mA.

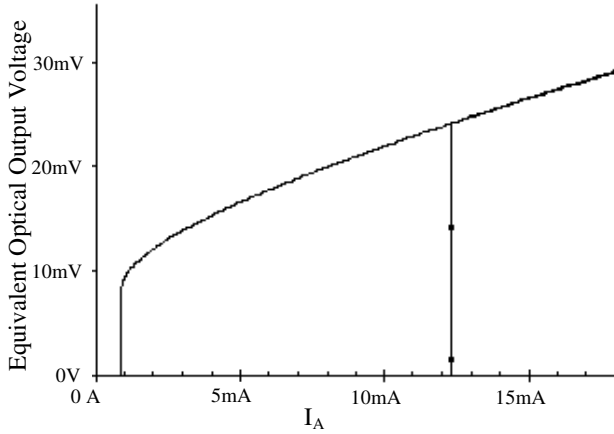


Fig.3. Bistability behavior of the HBSL

Table.1. Parameters used for the simulation study

	Description	Value
τ_{gs}	Carrier transport time across SCH region in gain region	45ps
τ_{as}	Carrier transport time across SCH region in absorber	45ps
τ_{gn}	Bimolecular recombination lifetime in gain region	1ns
τ_{an}	Bimolecular recombination lifetime in absorber region	1ns
τ_{atte}	Thermionic emission lifetime in absorber QW	5.6ps
N_w	Number of Well	3
τ_p	Photon Carrier lifetime	5.6e-12s
V_{gs}	Volume of SCH in gain region	22.5e-9cm ³
V_{as}	Volume of SCH in absorber	0.225e-12cm ³
V_{gQ}	Volume of QW in gain region	1.875e-12cm ³
V_{aQ}	Volume of QW in absorber region	1.875e-13cm ³
S_c	Normalized Constant for Simulation to run	1.3e18
ϵ	Gain compression factor	1e-17
β	Spontaneous emission factor	0.00001
T_{oa}	Characteristic temperature of active region	200K
R_{th}	Thermal resistance	360CW ⁻¹
T_D	Operating temperature	
C_{leak}	Leakage current dimensionless constant	1e-7
T_{ol}	Characteristic temperature of leakage current	27K
T_{th}	Thermal life time	1e-6s
α_g	Ratio of gain region length to cavity length	0.005
α_a	Ratio of absorber region length to cavity length	0.03

Fig. 4 shows that the hysteresis width and threshold currents are the function of the volumes of both gain and absorber regions. When volume of the gain region is increased, both forward and reverse threshold currents increase maintaining the hysteresis

width. Whereas, when the volume of the absorber region is changed, the reverse threshold current increases without affecting the forward threshold current thus, reducing the hysteresis width. This shows that the hysteresis width and threshold current depend on the physical dimensions of HBSL device. Threshold current also depends on the source resistance. It reduces when the source resistance of the gain region is increased maintaining the hysteresis width at almost the same level. However, the hysteresis width increases with the source resistance in absorber region (see Fig. 5). Fig 6 shows that I_{th} for the forward sweep remains unchanged in the temperature range of 273 -353 K; whereas, it increases for the reverse sweep. This reduces the hysteresis width at higher temperature.

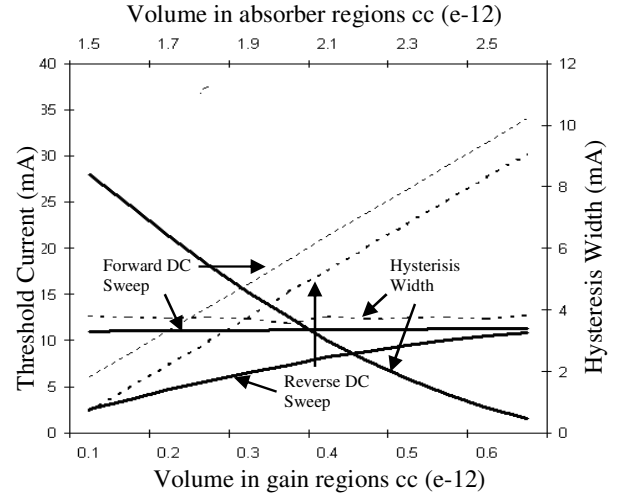


Fig. 4. Dependence of hysteresis width and threshold current on the volume of the gain region (dotted line) and the volume of the absorber region (solid line)

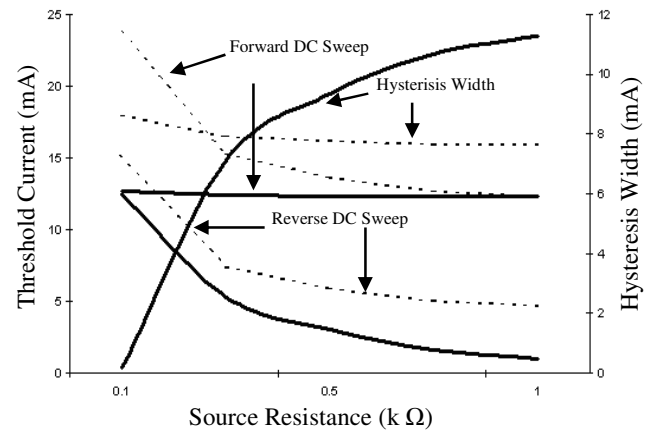


Fig. 5. Dependence of hysteresis width and threshold current on the source resistance of the gain region (dotted line) and that of absorber region (solid line)

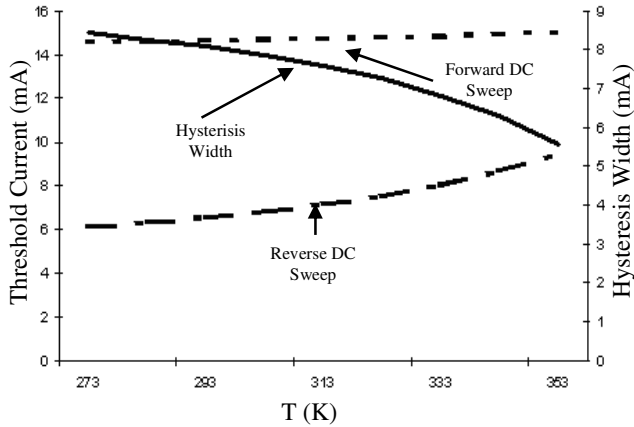


Fig.6. Variation of hysteresis width and threshold current with temperature

The transient response of the HBSL circuit model is simulated by biasing the gain region at 10 mA within the bistable region and applying positive pulse of 10mA height and 3ns width to the gain region. The device goes to the on-state and gives optical output. When a negative pulse of -10mA is applied, the optical output switches off as shown in Fig 7. We also noticed that the device shows a high turn-on delay due to the losses in the mirror (2 times longer than normal QWL). Turn-on delay can be reduced by applying a pulse of higher magnitude to the gain region (Fig 8). The optical output is found to increase with the input current.

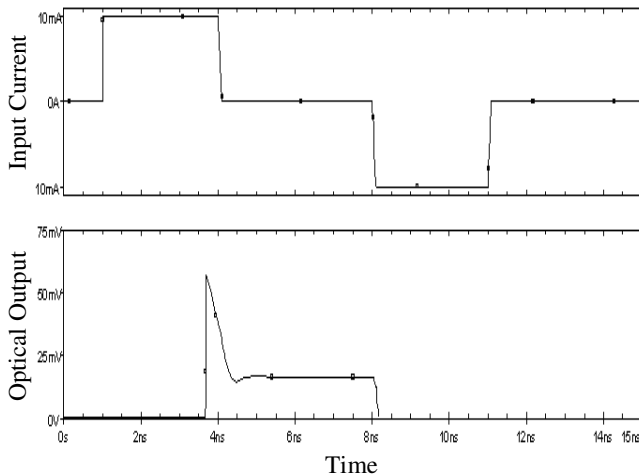


Fig.7. Transient Response of HBSL and the threshold current of hysteresis of the device

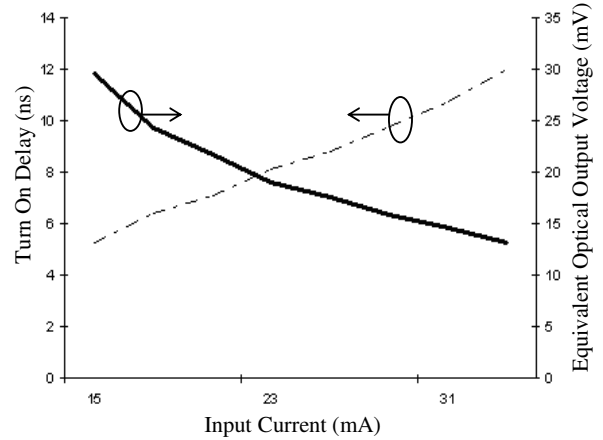


Fig.8. Effect of Input Current to the Transient Behaviour of the HBSL

5. CONCLUSION

In this paper, we have analyzed the hybrid semiconductor laser structure for optical bistability. The rate equations are transformed into equivalent electrical circuit. The circuit is stimulated using PSPICE to study the dc and transient behavior of the device. The dc sweep exhibits bistability. The hysteresis width and threshold current are found can be controlled by source resistance as well as by the device dimensions. The device exhibits long turn on delay, which can be improved by increasing the input current.

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